

Design of Wireless Energy-Harvested UHF WSN Tag for Cellular IoT

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Abstract

In this paper, a wireless energy-harvested ultra-high frequency (UHF) wireless sensor network (WSN) tag is designed and implemented for cellular IoT applications. The WSN tag is made up of a wireless energy harvesting circuit, a temperature sensing circuit, and a radio frequency identification (RFID) tag. The developed WSN tag is compatible with the ISO/IEC 18000-6C protocol. The WSN tag can receive the GSM RF energy operating in China GSM900 and GSM1800 bands in the surrounding environment and the solar energy, then converts the RF energy to direct current (DC) by schottky diode-based rectifying circuit, and finally stores the DC energy in a supercapacitor through a DC-DC booster circuit. The DC-DC booster circuit drives the front-end circuit, TI MSP430 microcontroller, temperature sensing circuit, and other active circuits in the tag. The MSP430 works in low-power mode when it is powered up, and it can also reduce power consumption more by reducing main clock (MCLK) frequency according to different forward link rates. The implemented WSN tag demonstrated that the RF-to-DC conversion efficiency is higher than 39% when the receiving 900 MHz RF signal power is from -14 dBm to 0 dBm and could make the tag work normally. The signal receiving sensitivity of the WSN tag is up to -32 dBm at the rate of 40 kbit/s from the Reader to the WSN tag. The WSN tag supports Miller coding and extended Miller coding. This wireless energy harvested UHF WSN tag, compared with conventional UHF passive tags and battery-powered active UHF RFID Tags, has many advantages, such as far communication distance, long service life, and sensing functionality. It will have wide applications in the Internet of Things (IoT).

Keywords

DC-DC booster circuit; MSP430; RF energy harvest; WSN tag

1 Introduction

Radio frequency identification (RFID) is a well-known non-contact automatic identification technology widely used in identification, supply chain, logistics, retail, manufacturing, garment industry, medical industry, identity and anti-counterfeiting and many other related applications [1]. RFID tags can be classified as passive tags, active tags and semi-active tags. The passive tags gain the power from reader emitting out RF signal in the near distance, so it has no DC power supply [2]. Active tags have

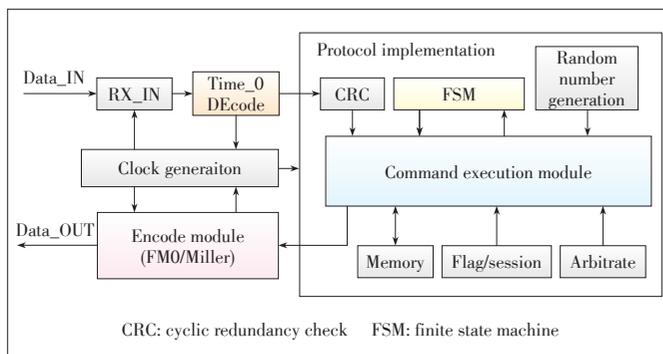
their own DC energy source and can transmit actively signal; therefore they have better performance than passive tags. Semi-active tags extract the power from their battery, and have a large operating distance.

In traditional wireless sensor network, almost all the wireless sensing nodes (it is also called as wireless sensing tags) operate in active mode or semi-active mode, they get the energy come from the rechargeable battery or other power supply. However, the rechargeable battery has the disadvantages of frequent maintenance and the limited number of charge cycles. For the advantage of battery-free operation, low-cost passive sensing tags have got much attention. A convenient solution is to provide energy for the sensing tags by harvesting ambient RF energy sources [3]. The RF energy from the environment can greatly extend the life of wireless sensor networks (WSN) [4]. The ambient RF energy is a type of new energy everywhere. Sustainability makes ambient RF energy more reliable than other types of new energy, such as solar energy and wind energy.

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▲ Figure 2. MCU achieving RFID tag protocol.

level and kept about $12.5 \mu\text{s}$. The MCU will execute Timer_0 module when the active delimiter is detected. The Timer_0 module achieves decoding function by timer interrupt, and the MCU will start working according to the command received. The Timer_0 module is configured as a rising edge trigger to receive the input signal. The MCU will trigger the interrupt when the rising edge is detected. The interrupt routine uses the current counter value minus the value of the previous counter value to indicate the time interval between two input signals at the rising edge. In the Timer_0 interrupt routine, the time interval of two adjacent rising edges will be frequently used to determine the level of input data. The Timer_0 module decodes the forward link data, extracts the forward link rate and $R_{\text{tcal}}/T_{\text{rcal}}$ length, and provides the relevant data for the execution module. If the downlink link rate is lower than 40 kbit/s, the MCLK frequency of the MCU can be set to lower for reducing power consumption according to the length of data0.

The command execution module implements all the other functions of the WSN tag, including cyclic redundancy check (CRC), finite state machine (FSM) implementation, random number generation, arbitration module, command processing, interrupt control, timing control, working mode switching, and so on. In this paper, the sensing command of detecting the ambient temperature and humidity has been implemented. The sensing functions are very important for a WSN tag; the tag collects various sensing information and replies the information to the reader, which may analyze the sensing information to get the specific environment information.

The random number generation module generates the 16 bits pseudo-random number for the communication reliability between tags and readers. The CRC16 of the $\{\text{PC} + \text{XPC} + \text{UII}\}$ calculates at power-up which is used as the seed of the pseudo-random number. The $\{\text{PC} + \text{XPC} + \text{UII}\}$ varies in different tags, so the seeds are different too. The pseudo-random number generation algorithm also provides the pseudo-random number for the handling signal. The encoding module encodes and transmits reply data. The MCU can configure a variety of clock frequency to support the different backscatter link frequency by calling the clock generation module. At the same time, the WSN tag supports the forward and backscatter rates less than

40 kbit/s and also supports extension miller subcarrier (the Miller sequence would contain exactly 2, 4, 8, 16, 32, 64, 128 or 256 subcarrier cycles per bit). The clock generation module may set the MCU clock cycles. The highest master clock frequency, MCLK, is 16 MHz. The MCU can quickly change its MCLK frequency through its registers base on the backscatter-link frequency (BLF) parameter.

3.1 Sensing Function

The custom sensing command is used to collect almost all sensory information, including temperature information, humidity information, pressure information, and other information. **Table 1** shows the formats of the custom sensing commands and **Table 2** shows its reply formats.

The command field is defined as 0xE001. The WSN field is a parameter of the custom sensing command, and its value represents the type of sensing information. That the WSN is 1 indicates the tag will collect temperature information, and that the WSN is 2 indicates the tag will collect humidity information.

The WrEn field indicates whether the WSN tag collects the sensing information or not, and whether it writes the information collected to memory or not. If WrEn is $2'b00$, it indicates that the WSN tag does not collect sensing information, reads the sensing information base on the address of the WSN tag distribution and returns sensing information to the reader. If WrEn is $2'b01$, it indicates that the WSN tag does not collect sensing information, reads the sensing information base on the address of the wordPtr field and returns sensing information to the reader. If WrEn is $2'b10$, it indicates that the WSN tag collects the sensing information, writes the sensing information to the assigned address by the WSN tag and returns the information to the reader. When WrEn is $2'b11$, it indicates that the WSN tag collects the sensing information, writes the sensing information to the address of WordPtr field and returns the sens-

▼ Table 1. The sensor command formats

	Bits	Description
Command	16	0xE001
WSN	16	0=RFU, 1=temperature, 2=humidity
RFU	16	0x0
WrEn	2	00=fix read, 01=read 10=fix write, 11=write
WordPtr	8	Word address pointer
RN16	16	Handle
CRC16	16	Checksum

▼ Table 2. Tag replies to the sensor commands

	Header	WSN Data	RN16	CRC16
Bits	1	16	16	16
Description	0	Data	Handle	Checksum

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ing information to the reader.

The WordPtr field indicates the address where the WSN tag writes the information collected to the memory. RN16 field is a 16-bit tag-authentication number. CRC16 field is the CRC check code for ensuring communication reliability. The custom sensing command has a total of 90 bits. The sensing information collected directly by the sensor connects to the analog to digital converter (ADC) of the MCU through analog input pin (the different sensing functions connect different input pins). ADC converts the input analog signal into the digital output (N_{ADC}) and stores N_{ADC} in the corresponding registers. The ADC supports 12-bit analog-to-digital conversion, and the ADC conversion formulas for N_{ADC} are shown as follows.

$$N_{ADC} = 4096 \times \frac{(V_{in+} + \frac{1}{2}LSB) - V_{R-}}{V_{R+} - V_{R-}}, \quad (2)$$

$$1LSB = \frac{V_{R+} - V_{R-}}{4096}. \quad (3)$$

Fig. 3 shows the execution of the custom sensing command. When receiving the sensing command, the MCU will jump to the subroutine and then verify the CRC-16 and 16-bit tag-authentication number (handle). If the CRC-16 and handle both pass the validation, the MCU will continue to execute this subroutine, otherwise it will terminate the process directly. The

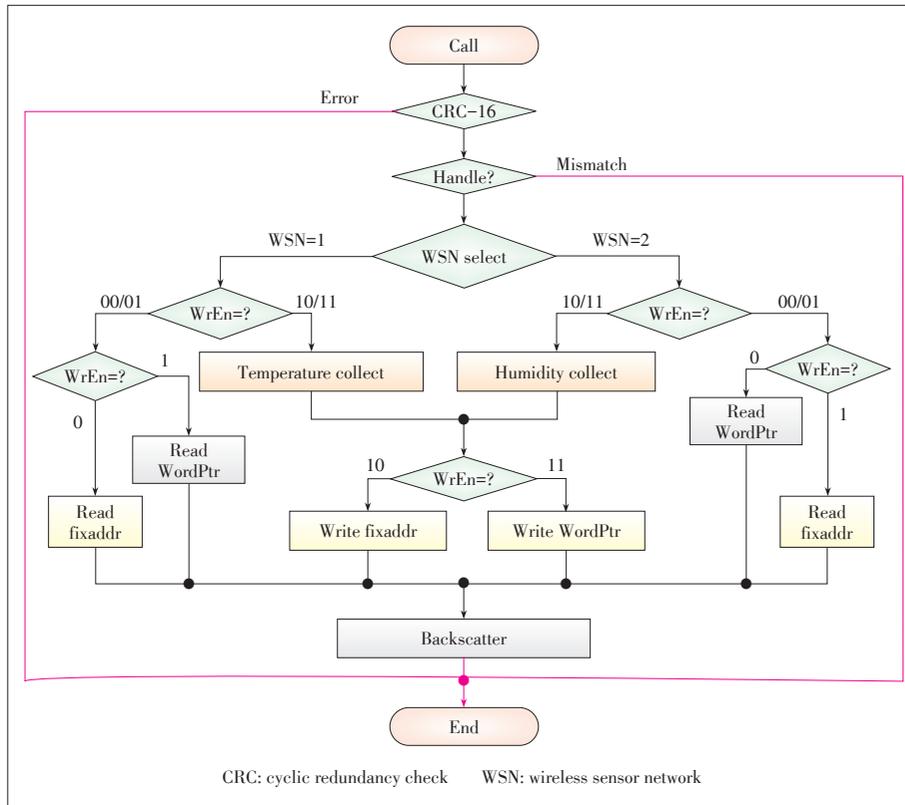
WSN field is used to select different functions of the wireless sensor network. The WrEn is used to select reading and writing. Finally, if the WSN tag passes the validation, it replies to the custom sensing command (Table 2), otherwise the WSN tag remains silent.

3.2 Miller Coding

The MCU realizes miller coding to meet the ISO/IEC18000-6C protocol standard. The miller coding rules are defined as follows. The original symbol “1” does not convert at the beginning and conducts conversion at the center point. The original symbol “0” is divided into a single “0” or a continuous “0” to be treated differently. When a single “0” appears, the electrical level before 0 is unchanged, even if the electrical level does not convert at the symbol boundary and is meta-intermediate. For the continuous “0”, the electrical level converts at the border of two consecutive “0”.

Firstly, the program encodes the miller preamble according to the TRext value of Query commands. When TRext is 0, the program sends the number of four times M values square waves; when TRext is 1, the program sends the number of sixteen times M values square waves. Secondly, the program adds to a string of 0100111_b at the end of square waves. The miller coding uses the shortest instruction to meet the timing requirement, and the jump range of the JMP instruction must be fewer than 512 words to avoid using extra clock cycles for the instruction. The MCU must separate two subcarrier sequences from other subcarrier sequences, because the two subcarrier sequences have only two subcarriers which are not possible to encode all bits as other subcarrier sequences encoding with five clock cycles. The loop jump needs to determine the number of subcarriers and the jump needs two clock cycles, but the program must determine to send 1 or 0 level in five clock cycles. Thus, it is necessary to distinguish two subcarrier sequences from other subcarrier sequences for meeting the timing requirement.

The miller coding is shown in Fig. 4. When the main program jumps to the miller coding through a jump command, the program must first configure the specific output ports and specific MCLK according to the BLF. Then, the program jumps to the M2 (two subcarriers) module when the M value specified in the Query command is one, otherwise it jumps to Mx (more subcarriers than two subcarriers) module. The TRext specified in the Query command determines the number of pilot tone sent. The program jumps to M_byte1_bit0 if the



▲ Figure 3. The execution of the sensing command.

bits_cnt (bits_cnt is zero for the entire byte, vice versa) is zero, otherwise jumps to Send_byte1_bit1. Taking miller value 2 as an example, Fig. 5 shows the encoded miller preamble.

The WSN tag expands the M values specified by the ISO/IEC18000-6C protocol standard, and the expanded M values shall be confirmed according to the custom command. For simplicity and compatibility, the target (Target = 101 and 110) of the select command is used to select different M values. The custom parameters are shown in Table 3.

4 Realization and Testing of Wireless Energy Harvest ER Ultra High Frequency IoT Tag

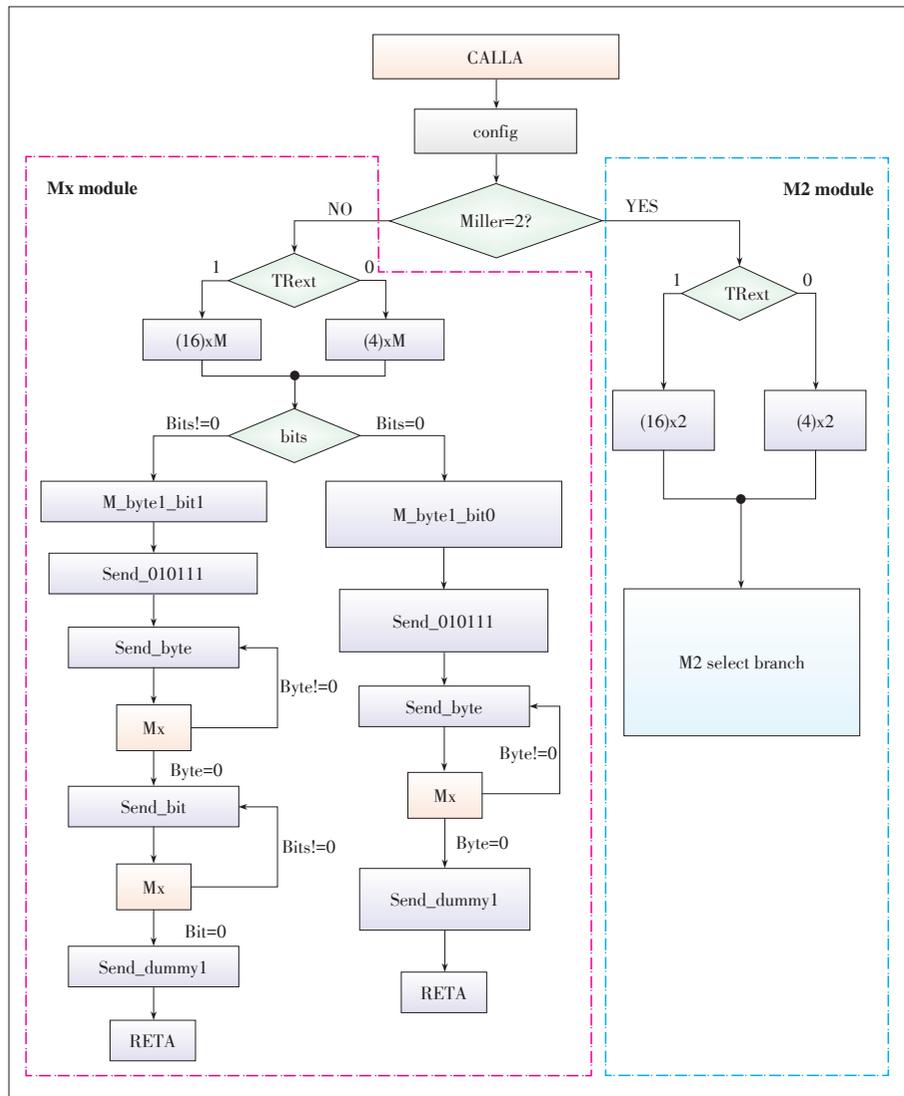
The proposed WSN tag was tested with a commercially available UHF RFID reader. The functions of the energy harvesting, ISO/IEC18000-6C protocol and temperature sensing were tested. The tested results are basically consistent with the theoretical analysis. The both front and back of the WSN tag circuit board are shown in Fig. 6.

4.1 Testing of Energy Harvesting

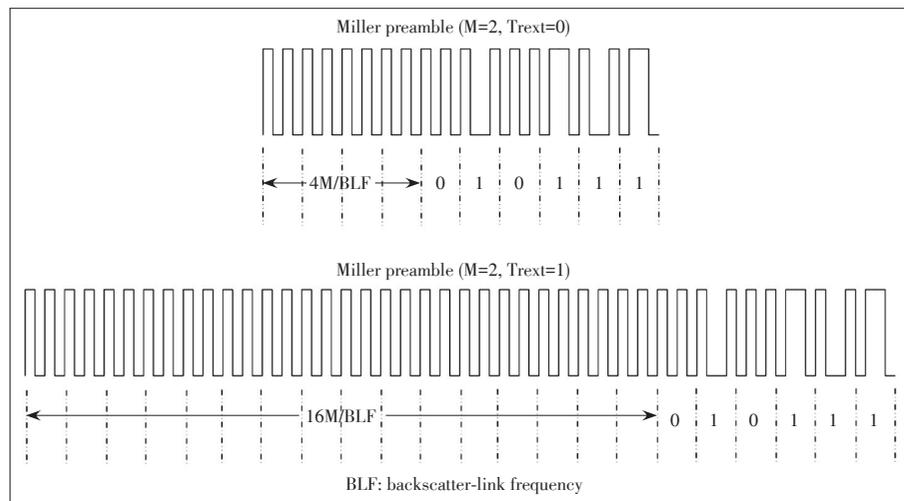
The WSN tag combines RF energy and solar energy for cellular IoT applications. The test of the energy harvesting performance was conducted in a wireless environment. The gain of transmitting antenna was 12 dBi, and the R420 reader provided 30 dBm transmit power. The gain of receiving antenna was 8 dBi. The two antennas kept 1 m away. The energy harvesting circuit can be charged in the wireless environment. The voltage of the super capacitor could reach 5.09 V within 24 s. The WSN tag can also turn solar energy into electricity by a solar panel, and the time of charging the supercapacitor up to 4.2 V is about 15 m.

4.2 Testing of Temperature Sensing

The temperature sensor was fixed to work in the mode which the gain selects are zero. The temperature sensing performance was tested in a high-low temperature test chamber from -30°C to 60°C with



▲ Figure 4. Miller coding.



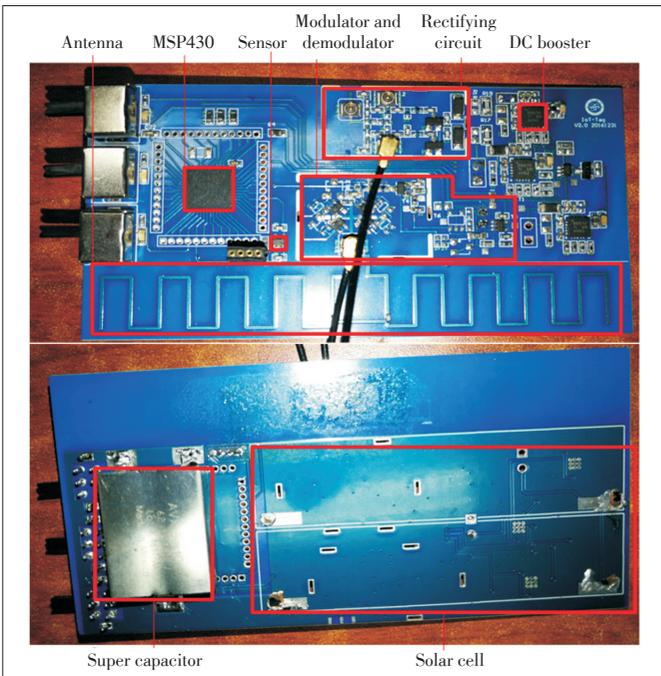
▲ Figure 5. The encoded miller preamble (M=2).

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▼Table 3. The compatibility table of the expanded M values

Target	M	Encode
3'bxxx	2'b00	FM0
3'b0xx	2'b01	M2
	2'b10	M4
	2'b11	M8
3'b101	2'b01	M16
	2'b10	M32
	2'b11	M64
3'b110	2'b01	M128
	2'b10	M256



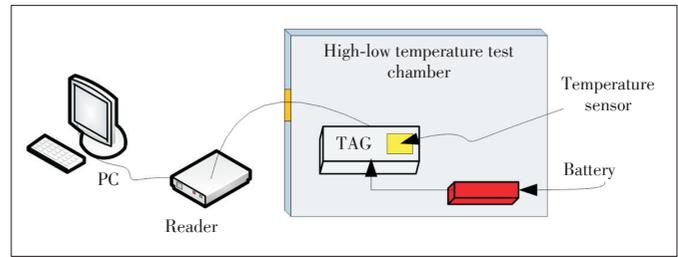
▲Figure 6. The circuit board of the WSN tag.

a step of 5°C. The test environment of WSN tag temperature is shown in Fig. 7. A self-designed UHF RFID reader was used for custom sensing command.

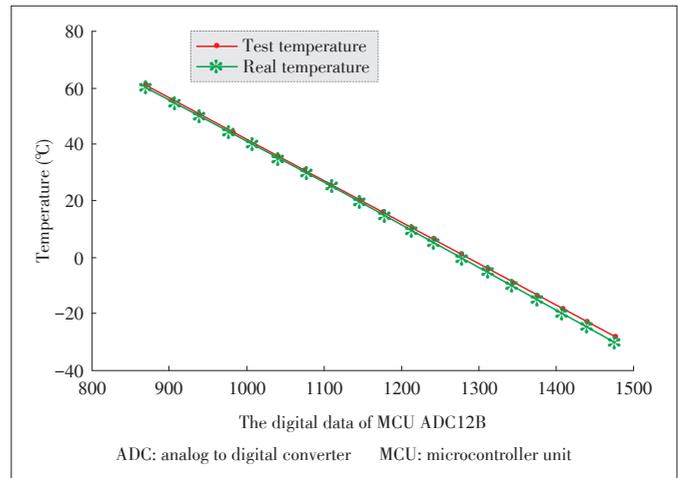
Fig. 8 compares the real temperature and test temperature. The horizontal axis represents N_{ADC} produced by the ADC, the vertical axis represents the temperature. We use the inverse transform of generating a N_{ADC} value to calculate the ambient temperature according to (1) and (2). From this figure, the test temperature is basically consistent with the real temperature.

4.3 Testing of Tag Protocol Performance

The distance within which the R420 reader can inventory the WSN tag successfully can be calculated by the friss electromagnetic wave propagation formula. The antennas of the reader and tag were placed on stands 2.4 m above the floor. The sensitivity of R420 reader is -82 dBm. The power transmitted



▲Figure 7. The test environment of WSN tag temperature.



▲Figure 8. The contrast of test temperature and real temperature.

is 30 dBm by the R420 reader, and the line loss L_{loss} is 1 dB. The gain of transmit antenna G_1 is 9 dBi, and the gain of tag antenna G_2 is 1 dBi. The sensitivity of tag, P_e , is -32 dBm, and the operating frequency is 920 MHz. The two antennas are both circular polarization, so the polarization loss, G_3 , is 0 dB. The loss of the backscatter, G_4 , is -6 dB. The distance within which the WSN tag can receive the available signal from the reader can be calculated by the following Friss equation:

$$P_e = P_1 - L_{loss} + G_1 + G_2 + G_3 + 10 \lg \left(\frac{\lambda}{4\pi r_1} \right)^2 \tag{4}$$

Therefore, $r_1 = 116$ m. Similarly, by the Friss equation, P_3 is the power reader received from the backscatter of the WSN tag. Accordingly, it is possible to calculate the distance r_2 within which the reader can receive available reply signals by (5).

$$P_3 = P_1 - 2L_{loss} + G_4 + 2(G_1 + G_2 + G_3) + 10 \lg \left(\frac{\lambda}{4\pi r_2} \right)^4 \tag{5}$$

Therefore, $r_2 = 32.68$ m. Table 4 shows the test distances in various modes of the reader, and some test distances are obviously beyond the theoretical distances, which are due to the influence of multipath effects.

5 Conclusions

The wireless energy-harvested WSN tag has been developed

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▼Table 4. Test distances in various modes

Reader mode	R->T (kbit/s)	T->R (kbit/s)	r_2 (m)	r_1 (m)
Autoset dense reader	≈70	≈426.66	19	91.2
Autoset static	≈70	≈426.66	20	93.8
Max throughput	≈70	≈426.66	23	90.9
Hybrid mode	≈70	≈426.66	27	94.9
Dense reader M=4(1)	≈50	≈320	33	96.4
Dense reader M=4(2)	≈70	≈320	35.2	96.4
Dense reader M=8	≈70	≈320	36.5	97.4

in circuit board (Fig. 6). Reducing power and obtaining renewable energy are main advantages of the WSN tag. This paper presents the relevant wireless energy harvesting, common tag protocol, miller coding, extended miller coding and sensing functions. The test results of energy harvesting are consistent with the theoretical values. The rectifier efficiency reaches 39% when the rectifier circuit has input of -14 dbm and output voltage of 360 mV that is greater than 330 mV, which meets the condition of the DC Booster circuit cold start. The WSN tag successfully implements the common ISO/IEC 18000-6C protocol standard, and the sensitivity of the WSN tag is up to -32 dBm. The WSN tag also implements low speed mode for saving power. The extensional miller coding has first been implemented in the assembler. In the future, we plan to design multi-band RF energy harvesting system, to implement more applications of the WSN tag for the cellular IoT.

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